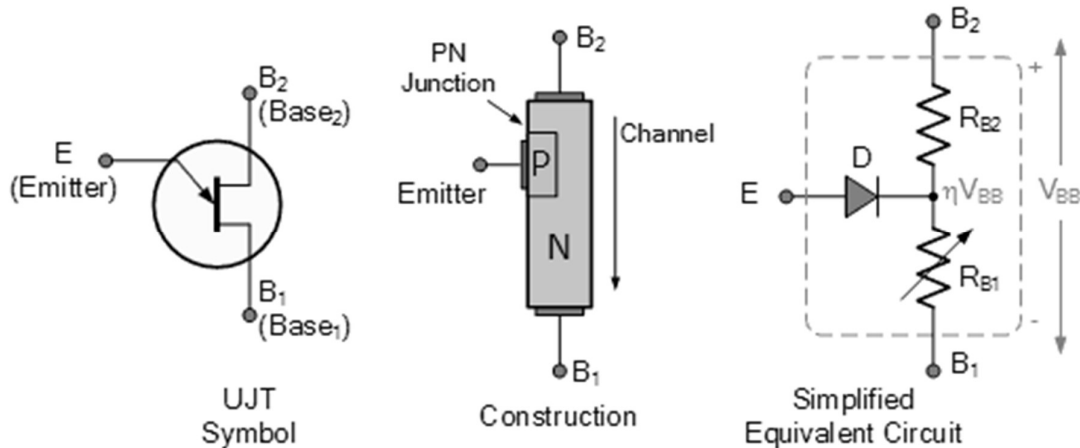


## UNIT V POWER DEVICES AND DISPLAY DEVICES

### 5.1 UNIJUNCTION TRANSISTOR (UJT)

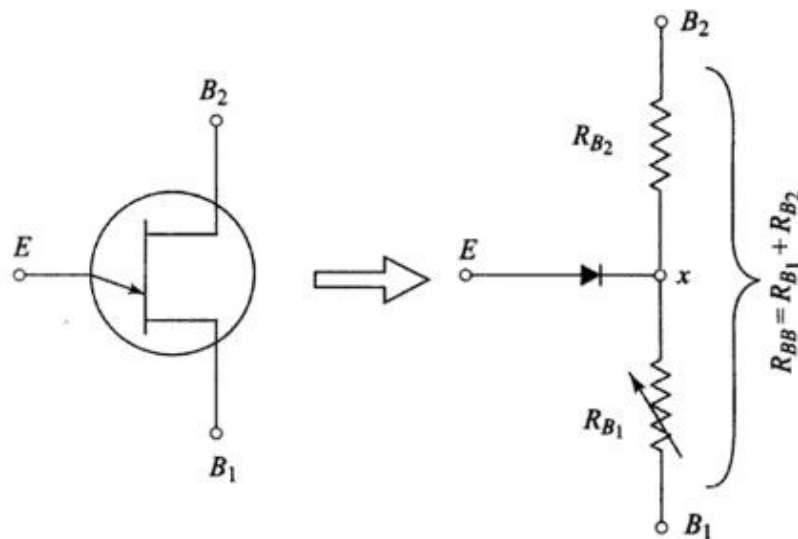
A typical UJT structure, pictured in figure, consists of a lightly doped, N-type silicon bar provided with ohmic contacts at each end. The two end connections are called base-1, designated B<sub>1</sub>, and base-2, B<sub>2</sub>. A small, heavily doped P-region is alloyed into one side of the bar closer to B<sub>2</sub>. This P-region is the UJT emitter E, and forms a P-N junction with the bar.



**Fig:5.1.1 (a) UJT Symbol, (b) UJT Construction, (c) Equivalent Circuit**

An interbase resistance,  $R_{BB}$ , exists between B<sub>1</sub> and B<sub>2</sub>. It is typically between 4 k $\Omega$  and 10k $\Omega$ , and can easily be measured with an ohmmeter with the emitter open.  $R_{BB}$  is essentially the resistance of the N-type bar. This interbase resistance can be broken up into two resistances, the resistance from B<sub>1</sub> to emitter, called  $R_{B1}$  and resistance from B<sub>2</sub> to emitter, called  $R_{B2}$ . Since the emitter is closer to B<sub>2</sub>, the value of  $R_{B1}$  is greater than  $R_{B2}$  (typically 4.2 k $\Omega$  versus 2.8 k $\Omega$ ).

The operation of the UJT can better be explained with the aid of an equivalent circuit. The UJT's circuit symbol and its equivalent circuit are shown in below. The diode represents the P-N junction between the emitter and the base-bar (point x). The arrow through  $R_{B1}$ , indicates that it is variable since during nonnal operation it may typically range from 4 k $\Omega$  down to 10  $\Omega$ .



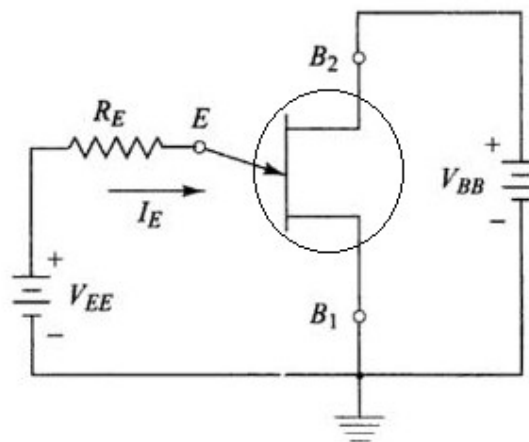
**Fig:5.1.2 UJT Equivalent Circuit**

### UJT operation

(a) When the emitter diode is reverse biased, only a very small emitter current flows. Under this condition,  $R_{B1}$  is at its normal high-value (typically 4 k $\Omega$ ). This is the UJT's "off" state.

(b) When the emitter diode becomes forward biased,  $R_{B1}$  drops to a very low value (reason to be explained later) so that the total resistance between E and B1 becomes very low, allowing emitter current to flow readily. This is the "on" state.

Circuit Operation of Uni junction Transistor (UJT)



**Fig:5.1.3 (a) Working of UJT**

The UJT is normally operated with both  $B_2$  and  $E$  biased positive relative to  $B_1$  as shown in below figure.  $B_1$  is always the UJT reference terminal and all voltages are measured relative to  $B_1$ . The  $V_{BB}$  source is generally fixed and provides a constant

voltage from B2 to B1. The VEE source is generally a variable voltage and is considered the input to the circuit. Very often, VEE is not a source but a voltage across a capacitor. The UJT circuit operation with the aid of the UJT equivalent circuit, shown inside the dotted lines in Fig.(a). We will also utilize the UJT emitter-base-1 VE-IE curve shown in Fig.(b). The curve represents the variation of emitter current IE, with emitter-base-1 voltage, VE, at a constant B2-B1 voltage. The important points on the curve are labelled, and typical values are given in parentheses.

The “Off” state If we neglect the diode for a moment, we can see in Fig.(a) that RB1 and RB2 form a voltage divider that produces a voltage Vx, from point x relative to ground.

$$V_x = \frac{R_{B_1}}{R_{B_1} + R_{B_2}} \times V_{BB} = \frac{R_{B_1}}{\underbrace{R_{BB}}_{\eta}} \times V_{BB}$$

Simply,

$$V_x = \eta V_{BB}$$

Where  $\eta$  (the greek letter “eta”) is the internal UJT voltage divider ratio RB1/RBB and is called the intrinsic stand of ratio.

Values of  $\eta$  typically range from 0.5 to 0.8 but are relatively constant for a given UJT.

The voltage at point x is the voltage on the N-side of the P-N junction. The VEE source is applied to the emitter which is the P-side. Thus, the emitter diode will be reverse-biased as long as VEE is less than Vx This is the “off” state and is shown on the VE-IE curve as being a very low current region. In the “off” state, then, we can say that the UJT has a very high resistance between E and B1, and IE is usually a negligible reverse leakage current. With no IE, the drop across RE is zero and the emitter voltage, VE, equals the source-voltage.

The UJT “off ” state, as shown on the VE-IE curve, actually extends to the point where the emitter voltage exceeds Vx by the diode threshold voltage, VD, which is needed to produce forward current through the diode. The emitter voltage and this point, P, is called the peak-point voltage, VP, and is given by

$$V_P = V_x + V_D = \eta V_{BB} + V_D$$

where  $V_D$  is typically 0.5 V. For example, if  $\eta = 0.65$  and  $V_{BB} = 20V$ , then  $V_P = 13.5 V$ . Clearly,  $V_P$  will vary as  $V_{BB}$  varies.

The "On" state As  $V_{EE}$  increases, the UJT stays "off" until  $V_E$  approaches the peak-point value  $V_P$ , then things begin to happen. As  $V_E$  approaches  $V_P$ , the P-N junction becomes forward biased and begins to conduct in the opposite direction.

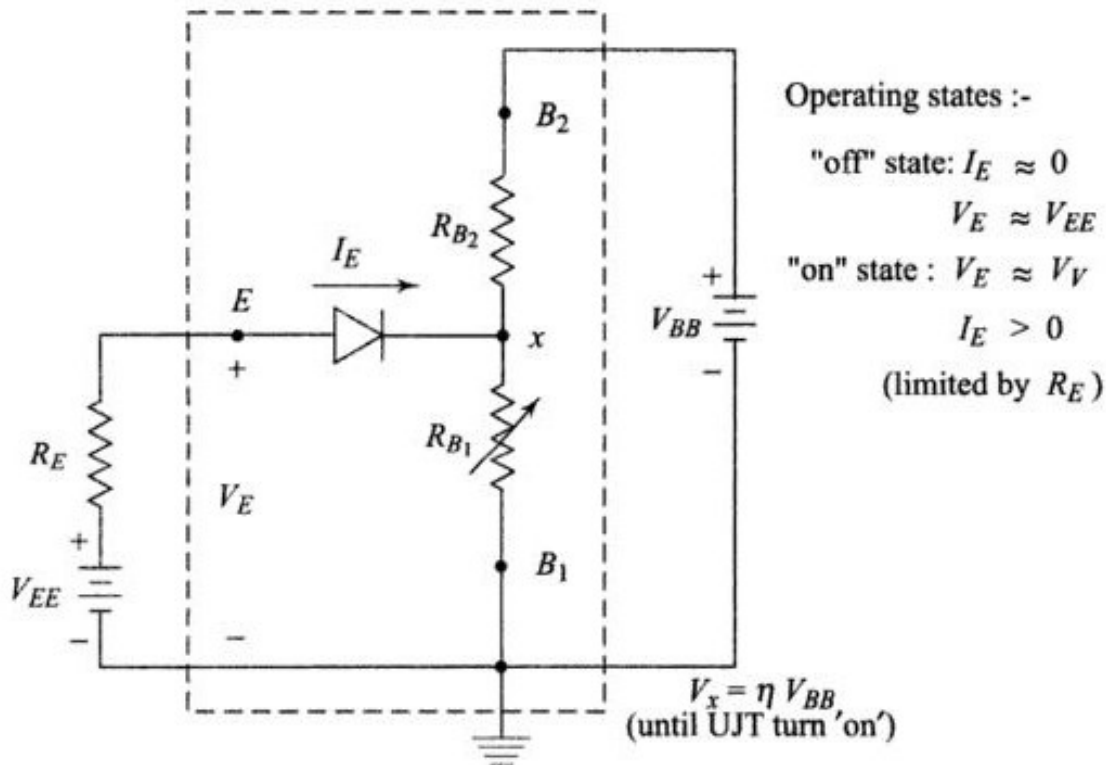
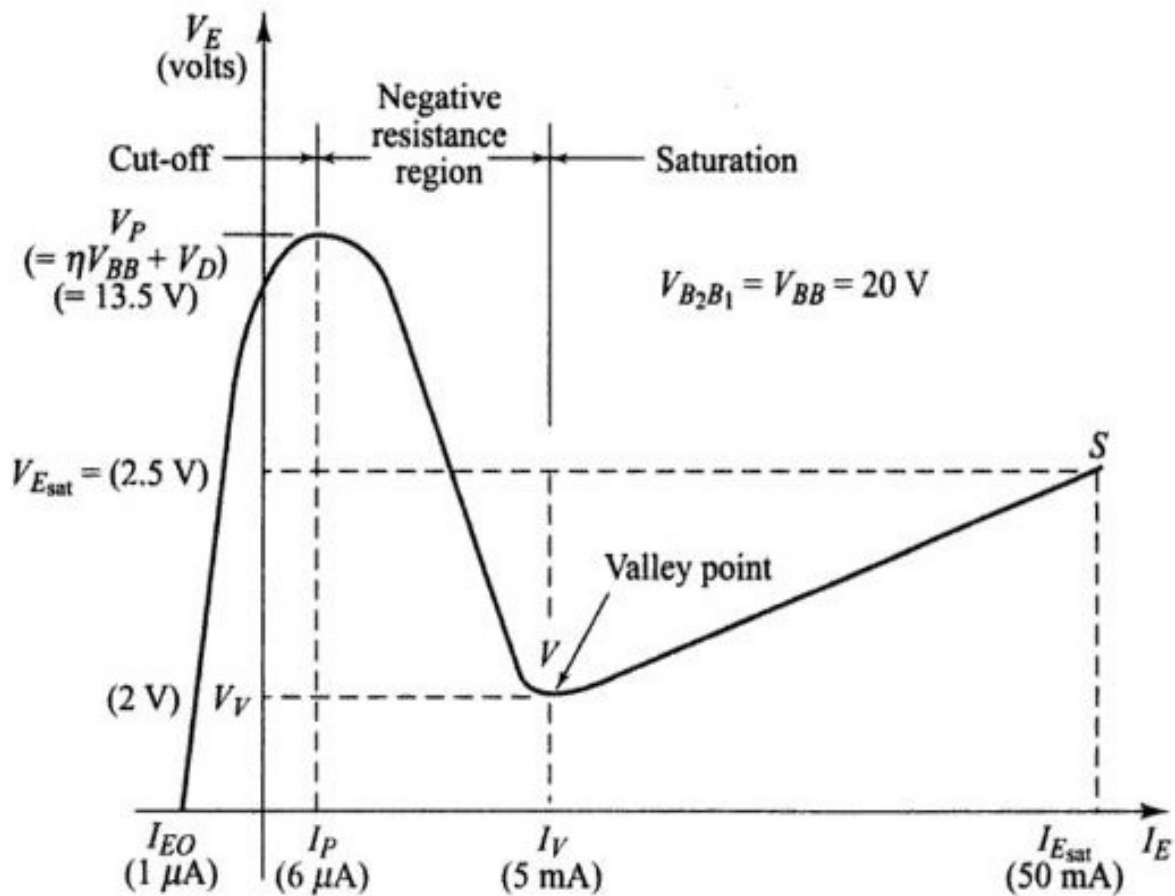


Fig:5.1.4 Equivalent Circuit for Analysis

Note on the  $V_E$ - $I_E$  curve that  $I_E$  becomes positive near the peak point P. When  $V_E$  exactly equals  $V_P$ , the emitter current equals  $I_P$ , the peak-point current. At this point, holes from the heavily doped emitter are injected into the N-type bar, specially into the B1 region. The bar, which is lightly doped, offers very little chance for these holes to recombine. As such, the lower half of the bar becomes replete with additional current carriers (holes) and its resistance  $R_{B1}$  is drastically reduced. The decrease in  $R_{B1}$  causes  $V_x$  to drop. This drop in turn causes the diode to become more forward biased, and  $I_E$  increases even further. The larger  $I_E$  injects more holes into B1 further reducing  $R_{B1}$ , and so on. When this regenerative or snowballing process ends,  $R_{B1}$  has dropped to a very small value (2-25  $\Omega$ ) and  $I_E$  can become very large, limited mainly by external resistance  $R_E$ .

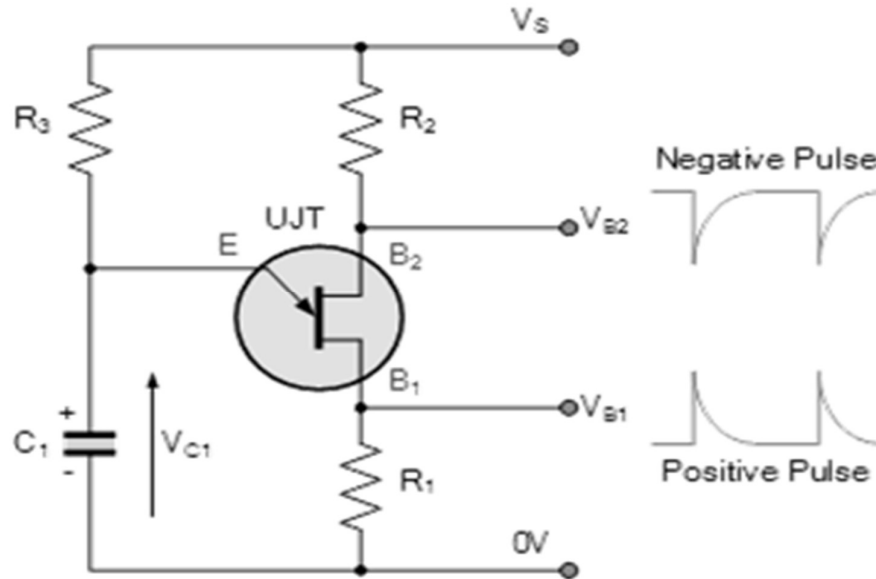


**Fig:5.1.5 VI Characteristics of UJT**

The UJT operation has switched to the low-voltage, high-current region of its  $V_E$ -  $I_E$  curve. The slope of this “on” region is very steep, indicating a low resistance. In this region, the emitter voltage  $V_E$ , will be relatively small, typically  $2V$ , and remains fairly constant as  $I_E$  is increased up to its maximum rated value,  $I_{E(sat)}$ . Thus, once the UJT is “on,” increasing  $V_{EE}$  will serve to increase  $I_E$  while  $V_E$  remains around  $2V$ .

Turning “Off” the UJT Once it is “on,” the UJT’s emitter current depends mainly on  $V_{EE}$  and  $R_E$ . As  $V_{EE}$  decreases,  $I_E$  will decrease along the “on” portion of the  $V_E$  -  $I_E$  curve. When  $I_E$  decreases to point  $V$ , the valley point, the emitter current is equal to  $I_V$ , the valley current, which is essentially the holding current needed to keep the UJT “on”. When  $V_{EE}$  is decreased below  $I_V$ , the UJT turns “off” and its operation rapidly switches back to the “off” region of its  $V_E$  -  $I_E$  curve, where  $I_E = 0$  and  $V_E = V_{EE}$ . The valley current is the counterpart of the holding current in PNP devices, and generally ranges between  $1$  and  $10$  mA.

## Unijunction Transistor Relaxation Oscillator



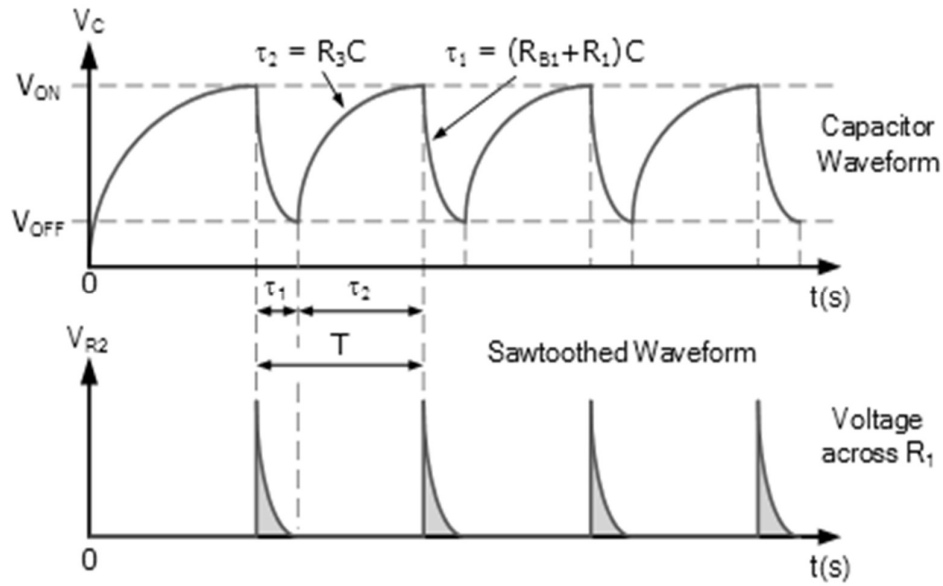
**Fig:5.1.6 UJT Relaxation Oscillator**

When a voltage ( $V_s$ ) is firstly applied, the unijunction transistor is “OFF” and the capacitor  $C_1$  is fully discharged but begins to charge up exponentially through resistor  $R_3$ . As the Emitter of the UJT is connected to the capacitor, when the charging voltage  $V_c$  across the capacitor becomes greater than the diode volt drop value, the p-n junction behaves as a normal diode and becomes forward biased triggering the UJT into conduction. The unijunction transistor is “ON”. At this point the Emitter to B1 impedance collapses as the Emitter goes into a low impedance saturated state with the flow of Emitter current through  $R_1$  taking place.

As the ohmic value of resistor  $R_1$  is very low, the capacitor discharges rapidly through the UJT and a fast rising voltage pulse appears across  $R_1$ . Also, because the capacitor discharges more quickly through the UJT than it does charging up through resistor  $R_3$ , the discharging time is a lot less than the charging time as the capacitor discharges through the low resistance UJT.

When the voltage across the capacitor decreases below the holding point of the p-n junction ( $V_{OFF}$ ), the UJT turns “OFF” and no current flows into the Emitter junction so once again the capacitor charges up through resistor  $R_3$  and this charging and discharging process between  $V_{ON}$  and  $V_{OFF}$  is constantly repeated while there is a supply voltage,  $V_s$  applied.

## UJT Oscillator Waveforms



**Fig:5.1.7 UJT Oscillator Waveforms**

Then we can see that the unijunction oscillator continually switches “ON” and “OFF” without any feedback. The frequency of operation of the oscillator is directly affected by the value of the charging resistance  $R_3$ , in series with the capacitor  $C_1$  and the value of  $\eta$ . The output pulse shape generated from the Base1 ( $B_1$ ) terminal is that of a sawtooth waveform and to regulate the time period, you only have to change the ohmic value of resistance,  $R_3$  since it sets the RC time constant for charging the capacitor.

The time period,  $T$  of the sawtoothed waveform will be given as the charging time plus the discharging time of the capacitor. As the discharge time,  $\tau_1$  is generally very short in comparison to the larger RC charging time,  $\tau_2$  the time period of oscillation is more or less equivalent to  $T \cong \tau_2$ . The frequency of oscillation is therefore given by  $f = 1/T$ .

### Applications of UJT:

Unijunction transistors are used extensively in oscillator, pulse and voltage sensing circuits. Some of the important applications of UJT are discussed below

- (i) UJT relaxation oscillator.
- (ii) Overvoltage detector.